$$\frac{1}{2\pi} \frac{d}{dt} \theta(t) = f(t) = f_{c} + k_{f} m(t)$$

5.2 FM and PM

Definition 5.15. Frequency modulation (FM):

$$x_{\rm FM}(t) = A \cos\left(2\pi f_c t + \phi + 2\pi k_f \int_{-\infty}^t m(\tau) d\tau\right).$$
(69)

Its instantaneous frequency is

$$f\left(t\right) = f_c + k_f m\left(t\right).$$

5.16. *Phase modulation* (PM): The phase-modulated signal is defined in Definition 5.3 to be O(t)

$$x_{\mathrm{PM}}(t) = A\cos\left(2\pi f_c t + \phi + k_p m\left(t\right)\right)$$

Its instantaneous frequency is

$$\mathcal{L}(t) = \frac{1}{2\pi} \frac{d}{dt} \theta(t) = \mathcal{L} + \frac{k_{P}}{2\pi} \frac{d}{dt} m(t)$$
(70)

Therefore, the instantaneous frequency of the PM signal varies in proportion with the slope of m(t).



Figure 32: A revisit of the PM signal in Figure 29. Analog modulation

(continuous ware)

Carrier :
$$A cos(2\pi f_c t + \emptyset)$$

AM : $A \rightarrow A + m(t)$
FM : $f(t) = f_c + k_F m(t)$
PM : $\beta \rightarrow \beta + k_F m(t)$
 $f(t) = f_c + \frac{k_P}{2\pi} m(t)$
 $discontinuous, have jump(s)$
the sudden change in the
phase is proportional to the
jump in m(t).

In particular, the instantaneous frequency of the PM signal is maximum when the slope of m(t) is maximum and minimum when the slope of m(t)is minimum.

Example 5.17. Sketch FM and PM waves for the modulating signal m(t) shown in Figure 33a.



Figure 33: FM and PM waveforms generated from the same message.

Example 5.18. Sketch FM and PM waves for the modulating signal m(t) shown in Figure 34a.



Figure 34: FM and PM waveforms generated from the same message.

The "indirect" method of sketching $x_{PM}(t)$ (using $\dot{m}(t)$ to frequencymodulate a carrier) works as long as m(t) is a continuous signal. If m(t)is discontinuous, this indirect method fails at points of discontinuities. In such a case, a direct approach should be used to specify the sudden phase changes.

5.19. Relationship between FM and PM:

- Equation (69) implies that one can produce frequency-modulated signal from a phase modulator.
- Equation (70) implies that one can produce phase-modulated signal from a frequency modulator.
- The two observations above are summarized in Figure 35.



Figure 35: With the help of integrating and differentiating networks, a phase modulator can produce frequency modulation and vice versa [4, Fig 5.2 p 255].

- By looking at an angle-modulated signal x(t), there is no way of telling whether it is FM or PM.
 - Compare Figure 29c and 29d in Example 5.6.
 - In fact, it is meaning less to ask an angle-modulated wave whether it is FM or PM. It is analogous to asking a married man with children whether he is a father or a son. [5, p 255]

5.20. Generalized angle modulation (or exponential modulation):

$$x(t) = A\cos\left(2\pi f_c t + \theta_0 + (m*h)(t)\right)$$

where h is causal.

(a) Frequency modulation (FM): $h(t) = 2\pi k_f 1[t \ge 0]$

(b) **Phase modulation** (\mathbf{PM}) : $h(t) = k_p \delta(t)$.

5.21. So far, we have spoken rather loosely of amplitude and phase modulation. If we modulate two real signals a(t) and $\phi(t)$ onto a cosine to produce the real signal $x(t) = a(t) \cos(\omega_c t + \phi(t))$, then this language seems unambiguous: we would say the respective signals amplitude- and phase-modulate the cosine. But is it really unambiguous?

The following example suggests that the question deserves thought.

Example 5.22. [8, p 15] Let's look at a "purely amplitude-modulated" signal

$$x_1(t) = a(t)\cos(\omega_c t).$$

Assuming that a(t) is bounded such that $0 \le a(t) \le A$, there is a welldefined function

$$\theta(t) = \cos^{-1}\left(\frac{1}{A}x_1(t)\right) - \omega_c t.$$

Observe that the signal

$$x_2(t) = A\cos\left(\omega_c t + \theta(t)\right)$$

is exactly the same as $x_1(t)$ but $x_2(t)$ looks like a "purely phase-modulated" signal.

5.23. Example 5.22 shows that, for a given real signal x(t), the factorization $x(t) = a(t) \cos(\omega_c t + \phi(t))$ is not unique. In fact, there is an infinite number of ways for x(t) to be factored into "amplitude" and "phase".